

BUH515D

HIGH VOLTAGE FAST-SWITCHING NPN POWER TRANSISTOR

- STMicroelectronics PREFERRED SALESTYPE
- HIGH VOLTAGE CAPABILITY
- U.L. RECOGNISED ISOWATT218 PACKAGE (U.L. FILE # E81734 (N))
- NPN TRANSISTOR WITH INTEGRATED FREEWHEELING DIODE

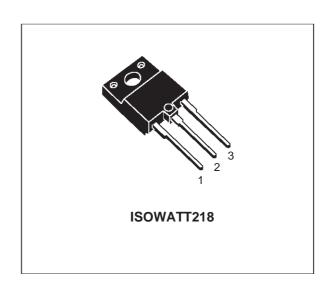
APPLICATIONS:

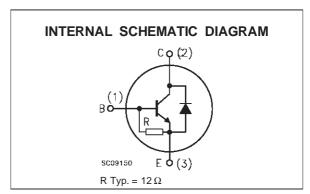
 HORIZONTAL DEFLECTION FOR COLOUR TV

DESCRIPTION

The BUH515D is manufactured using Multiepitaxial Mesa technology for cost-effective high performance and uses a Hollow Emitter structure to enhance switching speeds.

The BUH series is designed for use in horizontal deflection circuits in televisions and monitors.





ABSOLUTE MAXIMUM RATINGS

Symbol	Parameter	Value	Unit
V _{CBO}	Collector-Base Voltage (I _E = 0)	1500	V
V _{CEO}	Collector-Emitter Voltage (I _B = 0)	700	V
V _{EBO}	Emitter-Base Voltage (I _C = 0)	5	V
Ic	Collector Current	8	Α
I _{CM}	Collector Peak Current (tp < 5 ms)	15	Α
I _B	Base Current	5	Α
I _{BM}	Base Peak Current (t _p < 5 ms)	8	Α
P _{tot}	Total Dissipation at T _c = 25 °C	50	W
T _{stg}	Storage Temperature	-65 to 150	°C
Tj	Max. Operating Junction Temperature	150	°C

November 1999 1/7

BUH515D

THERMAL DATA

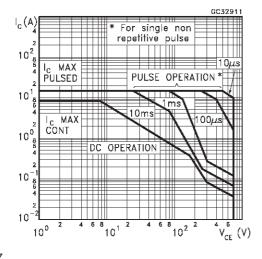
R _{thj-case} Thermal Resistance Junction-case	Max	2.5	°C/W
--	-----	-----	------

ELECTRICAL CHARACTERISTICS (T_{case} = 25 °C unless otherwise specified)

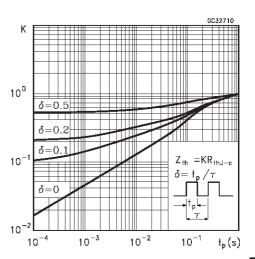
Symbol	Parameter	Test Conditions	Min.	Тур.	Max.	Unit
I _{CES}	Collector Cut-off Current (V _{BE} = 0)	$V_{CE} = 1300 \text{ V}$ $V_{CE} = 1500 \text{ V}$ $V_{CE} = 1500 \text{ V}$ $T_j = 125 \text{ °C}$			10 0.2 2	μA mA mA
I _{EBO}	Emitter Cut-off Current (I _C = 0)	V _{EB} = 5 V			200	mA
$V_{CE(sat)^*}$	Collector-Emitter Saturation Voltage	I _C = 5 A I _B = 1.25 A			1.5	V
V _{BE(sat)*}	Base-Emitter Saturation Voltage	I _C = 5 A I _B = 1.25 A			1.3	V
h _{FE} *	DC Current Gain	$I_{C} = 5 \text{ A}$ $V_{CE} = 5 \text{ V}$ $I_{C} = 5 \text{ A}$ $V_{CE} = 5 \text{ V}$ $T_{j} = 100 ^{\circ}\text{C}$	5 3		10	
t _s t _f	RESISTIVE LOAD Storage Time Fall Time	$V_{CC} = 400 \text{ V}$ $I_{C} = 5 \text{ A}$ $I_{B1} = 1.5 \text{ A}$ $I_{B2} = -2.5 \text{ A}$		2.4 170	3.6 260	μs ns
t _s t _f	INDUCTIVE LOAD Storage Time Fall Time	$\begin{aligned} &I_{C} = 5 \text{ A} & f = 15625 \text{ Hz} \\ &I_{B1} = 1.25 \text{ A} &I_{B2} = -2.5 \text{ A} \\ &V_{\text{ceflyback}} = 1050 \sin\!\left(\!\frac{\pi}{10}10^6\!\right)\!t & V \end{aligned}$		3.5 450		μs ns
VF	Diode Forward Voltage	I _F = 5 A			2	V

^{*} Pulsed: Pulse duration = 300 μs, duty cycle 1.5 %

Safe Operating Area

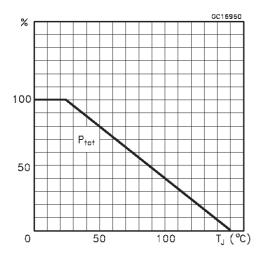


Thermal Impedance

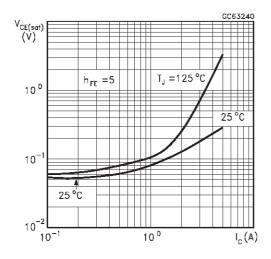


4

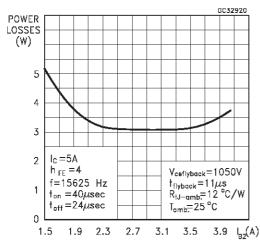
Derating Curve



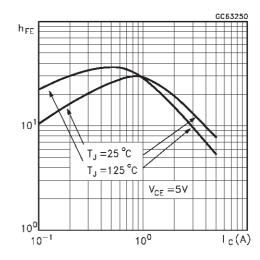
Collector Emitter Saturation Voltage



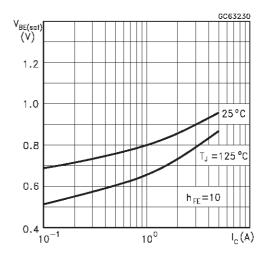
Power Losses at 16 KHz



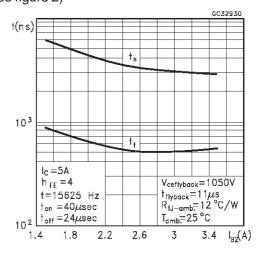
DC Current Gain



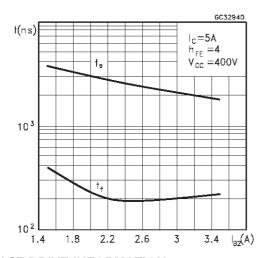
Base Emitter Saturation Voltage



Switching Time Inductive Load at 16KHz (see figure 2)



Switching Time Resistive Load



BASE DRIVE INFORMATION

In order to saturate the power switch and reduce conduction losses, adequate direct base current l_{B1} has to be provided for the lowest gain h_{FE} at $100\,^{\circ}C$ (line scan phase). On the other hand, negative base current l_{B2} must be provided to turn off the power transistor (retrace phase). Most of the dissipation, especially in the deflection application, occurs at switch-off. Therefore it is essential to determine the value of l_{B2} which minimizes power losses, fall time t_f and, consequently, T_j . A new set of curves have been defined to give total power losses, t_s and t_f as a function of l_{B2} at 16 KHz frequencies for choosing the optimum negative drive. The test circuit is illustrated in fig. 1.

Inductance L_1 serves to control the slope of the negative base current I_{B2} to recombine the excess carrier in the collector when base current is still present, this avoid any tailing phenomenon in the collector current.

The values of L and C are calculated from the following equations:

$$\frac{1}{2}L(I_C)^2 = \frac{1}{2}C(V_{CEfly})^2$$
$$\omega = 2\pi f = \frac{1}{\sqrt{LC}}$$

Where I_{C} = operating collector current, V_{CEfly} = flyback voltage, f= frequency of oscillation during retrace.

4/7

Figure 1: Inductive Load Switching Test Circuit

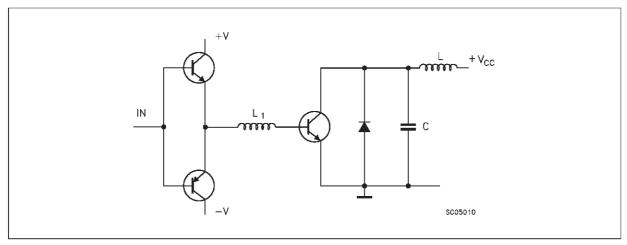
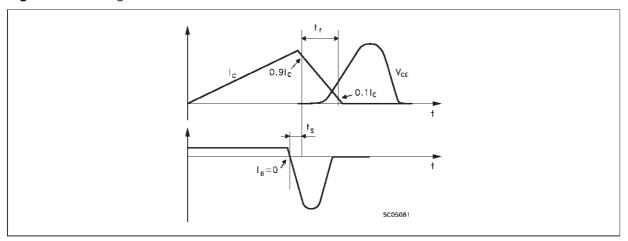
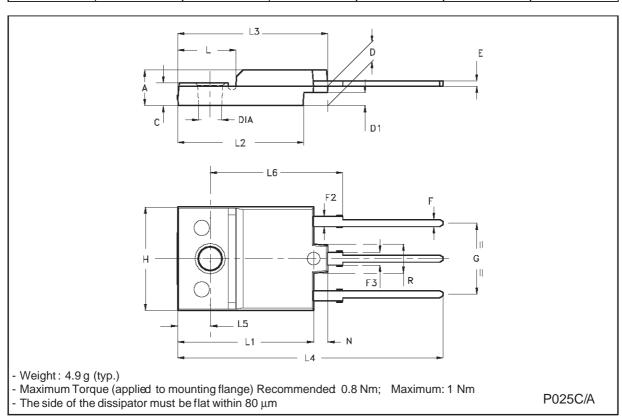


Figure 2: Switching Waveforms in a Deflection Circuit



ISOWATT218 MECHANICAL DATA

DIM.	mm			inch		
DIIVI.	MIN.	TYP.	MAX.	MIN.	TYP.	MAX.
Α	5.35		5.65	0.211		0.222
С	3.30		3.80	0.130		0.150
D	2.90		3.10	0.114		0.122
D1	1.88		2.08	0.074		0.082
Е	0.75		0.95	0.030		0.037
F	1.05		1.25	0.041		0.049
F2	1.50		1.70	0.059		0.067
F3	1.90		2.10	0.075		0.083
G	10.80		11.20	0.425		0.441
Н	15.80		16.20	0.622		0.638
L		9			0.354	
L1	20.80		21.20	0.819		0.835
L2	19.10		19.90	0.752		0.783
L3	22.80		23.60	0.898		0.929
L4	40.50		42.50	1.594		1.673
L5	4.85		5.25	0.191		0.207
L6	20.25		20.75	0.797		0.817
N	2.1		2.3	0.083		0.091
R		4.6			0.181	
DIA	3.5		3.7	0.138		0.146



6/7

Information furnished is believed to be accurate and reliable. However, STMicroelectronics assumes no responsibility for the consequences of use of such information nor for any infringement of patents or other rights of third parties which may result from its use. No license is granted by implication or otherwise under any patent or patent rights of STMicroelectronics. Specification mentioned in this publication are subject to change without notice. This publication supersedes and replaces all information previously supplied. STMicroelectronics products are not authorized for use as critical components in life support devices or systems without express written approval of STMicroelectronics.

The ST logo is a trademark of STMicroelectronics

© 1999 STMicroelectronics – Printed in Italy – All Rights Reserved STMicroelectronics GROUP OF COMPANIES

Australia - Brazil - China - Finland - France - Germany - Hong Kong - India - Italy - Japan - Malaysia - Malta - Morocco - Singapore - Spain - Sweden - Switzerland - United Kingdom - U.S.A.

http://www.st.com

